



## Intelligent Speed Controller Technique Based on Optimized Fuzzy Gains

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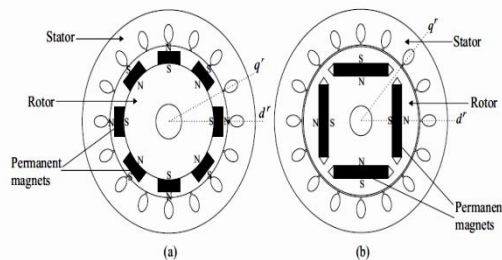
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**Abstract:** This article proposes the solution for high-performance applications, the Permanent Magnet Synchronous Motors (PMSMs) gained the attraction as compared to other types of ac motors due to some of their advantageous characteristics including high torque, high power, high efficiency, and low noise. Insensitivity to parameter variation and, reaching the speed to a reference value at the shortest time due to any disturbances, are some of the important criteria of the high-performance drive systems used for drive PMSM. The classical proportional-integral (PI) and proportional integral derivative (PID) controllers have been widely employed as speed controllers in PMSM controller's drives. In this work, a Fuzzy logic Controller for controlling the high speed of the Permanent Magnet Synchronous Motor is proposed to achieve a stable state. The fuzzy logic controller is more efficient and gives fast response than the other conventional controller. The simulation results show that out of three controllers, the fuzzy logic controller outperforms the other for the speed control of permanent magnet synchronous drives.

**Index Terms - Fuzzy Logic Controller, PI Controller, PID**

### I. INTRODUCTION

A permanent magnet synchronous motor (PMSM) is a motor that uses permanent magnets to produce the air gap magnetic field rather than using electromagnets. These motors have significant advantages, attracting the interest of researchers and industry for use in many applications. In recent years, permanent magnet synchronous motor drives have been widely used in many industrial applications such as robots, rolling mills, and machine tools. The inherent advantages of these machines include high power density, low inertia, and high-speed capabilities [1-5]. These aspects make PMSM performance surpasses that of the conventional dc and other ac motors in drives, especially where the overall efficiency is critical. As a result, PMSM drives become competitive to other drives in modern industrial applications requiring precision, efficient and sophisticated products, and services. Traditionally commutator motors, also known as direct current (dc) motors were preferred for variable However, the ac motor control including control of PMS motors is a challenging task due to very fast motor dynamics and highly non-linear models of the machines. Therefore, a major part of motor control development consists of deriving mathematical models in suitable forms. The dynamic models of the motors can be presented in different reference frames to lay down a basis for the motor control design. The mathematical formulations and the equivalent circuit models can be provided to help in better controller design for PMSM drives. There are two competing control strategies for ac motors i.e. vector control (VC) and direct torque control (DTC) for PMSM [8], [9],[10].

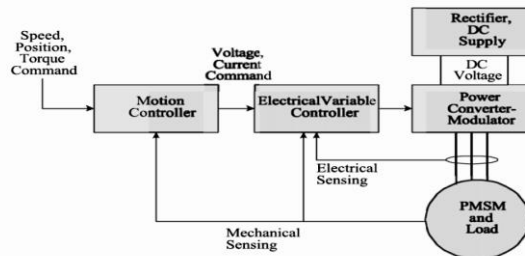


**Fig. 1** Motor cross-sections showing different rotor configuration for PMSMs. (a) Surface magnets type (b) Interior magnets type.

The interior magnets type rotor configuration brings saliency characteristics to the machine which is not present in a machine with surface magnets type rotor. As shown in fig.1 (a) and fig.1 (b), the magnetic flux induced by the magnets define the rotor director dr -axis radial through the centerline of the magnets. The rotor quadrature or qr -axis is orthogonally (90 electrical degrees) placed with the rotor dr -axis. Since the permeability of permanent-magnets is almost the same as the air, in\ interior magnets type configuration the effective airgap of dr -axis is increased compared to the q r -axis. Therefore, the dr -axis reluctance is higher than the q r -axis reluctance. This results in the q r -axis inductance is higher than the dr -axis inductance, i.e.  $L_q > L_d$ , in IPMSMs [18],[19].

**II. DRIVE SYSTEM OF PMSM**

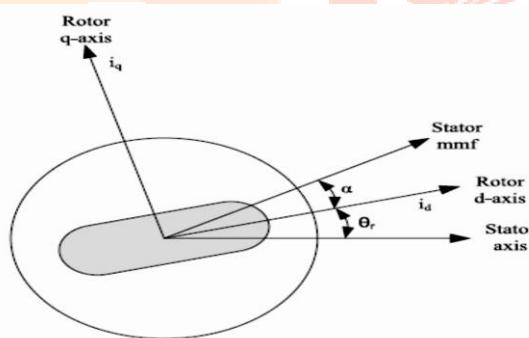
A conceptual drive system is pictured in fig 2. There, a speed, position, or torque command is input to the drive system. The motion controller implements feedback control based on mechanical sensors (or estimators). The controller outputs a command for the electrical variables to obey. The electrical control block converts its input commands into commands for the power converter/modulator block and sometimes utilizes feedback of voltage or current. The power converter block imposes the desired electrical signals onto the PMSM machine with the connected load [33], [36].



**Fig. 2** Diagram of conceptual drive system

**III. DETAILED MODELING OF PMSM**

Detailed modeling of the PM motor drive system is required for proper simulation of the system. The d-q model has been developed on the rotor reference frame as shown in Fig.3. At any time t, the rotating rotor d-axis makes an angle  $\theta_r$  with the fixed stator phase axis and rotating stator MMF makes an angle  $\alpha$  with the rotor d-axis. Stator MMF rotates at the same speed as that of the rotor.



**Fig. 3** Motor axis

The model of PMSM without damper winding has been developed on rotor reference frame using the following assumptions:

- 1) Saturation is neglected.
- 2) The induced EMF is sinusoidal.
- 3) Eddy currents and hysteresis losses are negligible.
- 4) There are no field current dynamics.

Voltage equations are given by:

$$V_q = R_s i_q + \omega_r \lambda_d + \rho \lambda_q \tag{1}$$

$$V_d = R_s i_d - \omega_r \lambda_q + \rho \lambda_d \tag{2}$$

Flux Linkages are given by

$$\lambda_q = L_q i_q \tag{3}$$

$$\lambda_d = L_d i_d + \lambda_f \tag{4}$$

Substituting equations 3.3 and 3.4 into 3.1 and 3.2

$$V_q = R_s i_q + \omega_r (L_d i_d + \lambda_f) + \rho L_q i_q \tag{5}$$

$$V_d = R_s i_d - \omega_r L_q i_q + \rho (L_d i_d + \lambda_f) \tag{6}$$

Arranging equations in matrix form

$$\begin{pmatrix} V_q \\ V_d \end{pmatrix} = \begin{pmatrix} R_s + \rho L_q & \omega_r L_d \\ -\omega_r L_q & R_s + \rho L_d \end{pmatrix} \begin{pmatrix} i_q \\ i_d \end{pmatrix} + \begin{pmatrix} \omega_r \lambda_f \\ \rho \lambda_f \end{pmatrix}$$

The developed torque motor is being given by

$$T_e = (3/2)(P/2)X(\lambda_d i_q - \lambda_q i_d)$$

The mechanical Torque equation is

$$T_e = T_L + B\omega_m + j\left(\frac{d\omega_m}{dt}\right) \tag{7}$$

Solving for the rotor mechanical speed form equation

$$\omega_m = \int \left( \frac{T_e - T_L - B\omega_m}{J} \right) dt \tag{8}$$

$$\text{and } \omega_m = \omega_r \left( \frac{2}{p} \right) \tag{9}$$

In the above equations,  $\omega_r$  is the rotor electrical speed whereas  $\omega_m$  is the rotor mechanical speed.

**IV. MATHEMATICAL MODEL OF PMSM**

The motor currents are decomposed into  $i_d$  and  $i_q$  components which are respectively flux and torque components in the rotor based d-q coordinates system. Motor model is constituted with following equations:

$$T_e = \frac{3}{2} \frac{P}{2} [\lambda_m \cdot i_q + (L_d - L_q) i_d \cdot i_q] \tag{10}$$

$$d(i_d)/dt = (v_d - r_s \cdot i_d + w_r \cdot L_q \cdot i_q) / L_d \tag{11}$$

$$d(i_q)/dt = [v_q - r_s \cdot i_s - w_r \cdot (L_d \cdot i_d + \lambda_m)] / L_q \tag{12}$$

$$d(\omega_r)/dt = (T_e - T_L - B \cdot \omega_{rm}) / J \tag{13}$$

$$\omega_r = \frac{P}{2} X \omega_r \tag{14}$$

where,

$T_L$  = load torque,

$B$  = viscous friction,

$J$  = moment of inertia,

$V_d$  &  $V_q$  = represent the d-q axis stator voltages,

$I_d$  &  $i_q$  = the d-q axis stator currents.

$L_d$  &  $L_q$  = the d-q axis inductances,

$r_s$  = per phase stator resistance,

$\omega_r$  = velocity of the rotor.

$\lambda_m$  = expression of the flux linkage due to the rotor magnets linking the stator,

$T_e$  = motor produced torque

$\omega_{rm}$  = mechanical velocity of the rotor.

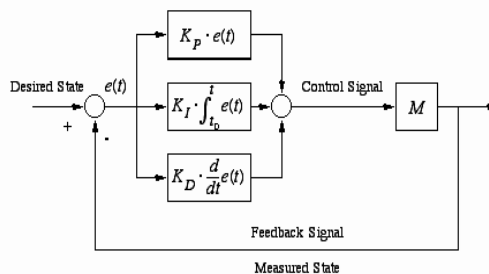
**V. Conventional Controller**

The most common controller action used in process control is one or a combination of continuous control modes and the method by which the controller counteracts a deviation from a setpoint is called a control mode. The three most commonly used modes of feedback control are proportional, integral, and derivative. Various combinations of these modes comprise mostly of the controllers found in industries. The various combinations generally found are:

PI (Proportional Integral) mode

PD (Proportional derivative) mode

PID (Proportional Integral Derivative) mode



VI. Fig -4: PID controller

The adoption of improved control algorithms has, however, been slow. Many computer control implementations have simply taken over the well-established analog three-term control algorithm (proportional + integral + derivative – PID). This has the general form

$$m = K_p (e + 1/T_i \int e \cdot dt + T_d \cdot de/dt) \tag{15}$$

In ‘proportional control’, actuating signal at the output of the controller is related to the input of the controller by a proportional constant. A linear continuous-data controller should be also able to take a time derivative or a time integral of the input signal, in addition to the proportional and other simple algebraic operations, such as addition and subtraction. One of the best-known controllers used is the PID controller where the PID stands for Proportional, Integral, and Differential. The transfer function of PID is:

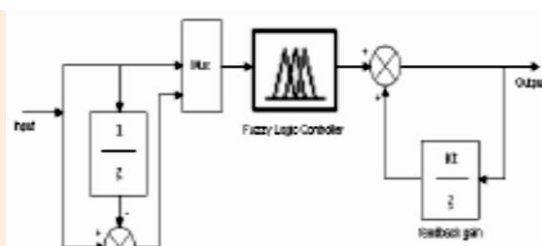
$$G(s) = Kp + \frac{ki}{s} + Kds \tag{16}$$

Where  $e = r - y$  and  $y$  is measured variable,  $r$  is reference value or setpoint, and  $e$  is error;  $Kp$  is the overall controller gain;  $Ti$  is the integral action time, and  $Td$  is the derivative action time. This algorithm can be expressed in another form. For example, derivative action is frequently not used, when  $de/dt$  is often replaced by  $dy/dt$  to avoid differentiating the set point.

The algorithm is normally implemented in DDC systems by using the difference equation equivalent to 1 above. If the sampling interval for computation is  $T$  seconds then the simple approximations.

**VII. Fuzzy Logic Controller**

Fuzzy logic is a thinking process or problem-solving control methodology incorporated in control system engineering, to control systems when inputs are either imprecise or the mathematical models are not present at all. Fuzzification is the process of making a crisp quantity into the fuzzy [18]. They carry considerable uncertainty. If the form of uncertainty happens to arise because of imprecision, ambiguity, or vagueness, then the variable is probably fuzzy and can be represented by a membership function. Defuzzification is the conversion of a fuzzy quantity to a crisp quantity, just as fuzzification is the conversion of a precise quantity to a fuzzy quantity [19]. There are many methods of defuzzification, out of which the smallest of the maximum method is applied in making a fuzzy inference system. The Fuzzy logic control consists of three main stages, namely the fuzzification interface, the inference rules engine, and the defuzzification interface. For speed control, the process operator is assumed to respond to variables error ( $e$ ) and change of error ( $ce$ ). The fuzzy logic controller with error and change in error is shown in fig. 5.

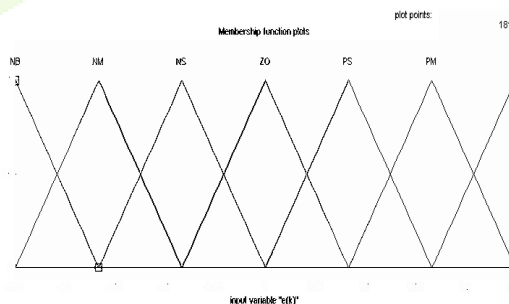


**Fig. 5 Fuzzy Logic Controller**

Taking the scaling gains into account, the global function of the FLC output signal can be written as.

$$\Delta Pc = F[n_c e(k), n_{ce} ce(k)] \tag{17}$$

Where  $n_e$  and  $n_{ce}$  are the error and error rate scaling gains, respectively, and  $F$  is a fuzzy nonlinear function. FLC is dependent on its inputs scaling gains [15]-[18],[19]. A label set corresponding to linguistic variables of the input control signals,  $e(k)$  and  $ce(k)$ , with a sampling time of 0.01 sec is given Attempt has been made to examine with Seven number triangular membership function (MFs) namely Negative Big(NB), Negative Medium(NM), Negative Small(NS), Zero(ZO), Positive Small(PS), Positive Medium(PM) and Positive Big(PB) are used. The numbers of rules are 49. The membership functions (MFs) for the input variables are shown in Fig.6.



**Fig. 6 Membership Function for the control input variables**

**Table 1**Fuzzy Inference Rule for Fuzzy Logic Controller

Input	e(k)							
		NB	NM	NS	ZO	PS	PM	PB
ce(k)	NB	PB	PB	PB	PB	PM	PM	PS
	NM	PB	PM	PM	PM	PS	PS	PS
	NS	PM	PM	PS	PS	PS	PS	ZO
	ZO	NS	NS	NS	ZO	PS	PS	PS
	PS	ZO	NS	NS	NS	NS	NM	NM
	PM	NS	NS	NM	NM	NM	NB	NB
	PB	NS	NM	NB	NB	NB	NB	NB

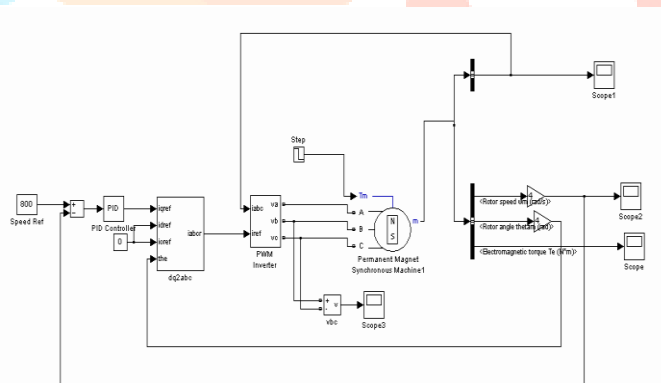
**IV. RESULTS AND DISCUSSION**

Speed controlling of permanent magnet synchronous motor has been observed using simulink software through different controllers i.e. using PID controller and Fuzzy logic controller,

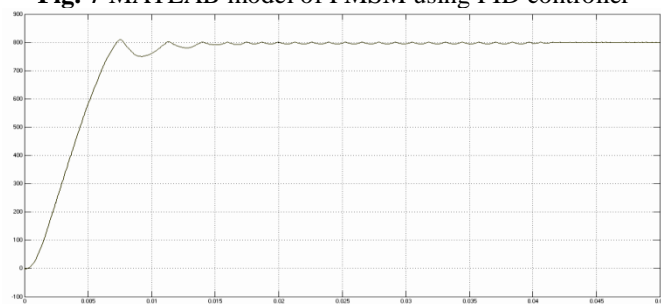
Table 2: Permanent Magnet Synchronous Motor parameters

Mechanical input	Torque
Stator resistance Rs(ohm)	2.85
Inductances Ld(H) Lq(H)	8.5e-3, 8.5e-3
Flux induced by magnets(Wb)	.175
Ref Speed	800 RPM

**7.1 Speed control model using PI and PID controller**



**Fig. 7** MATLAB model of PMSM using PID controller



**Fig. 8** Scope view of the speed of PMSM using a PI controller

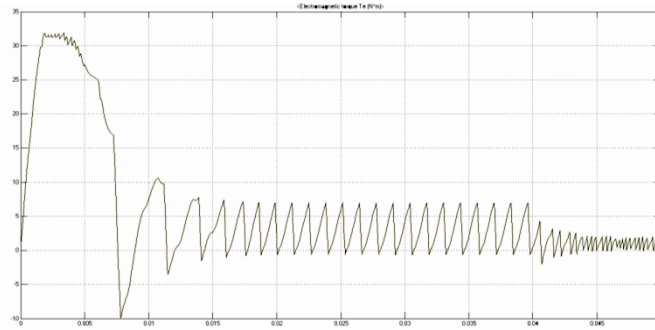


Fig. 9 Scope view of torque of PMSM using the PI controller

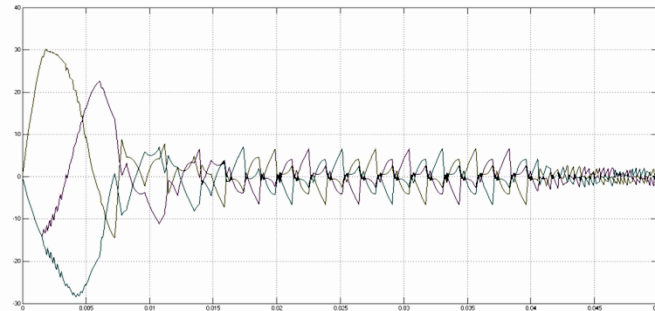


Fig. 10 Scope view of 3 phase current of PMSM using the PI controller

From the above model with PI controller and its response, we can see that the rising time and settling time with dampings comes after a long period so here we can say that controlling of permanent magnet synchronous motor is difficult using a PI controller.

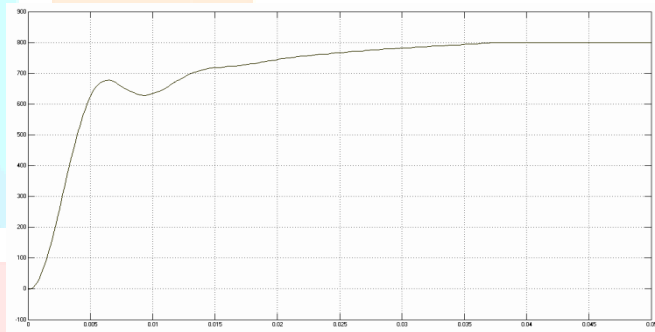


Fig. 11 Scope view of the speed of PMSM with PID controller

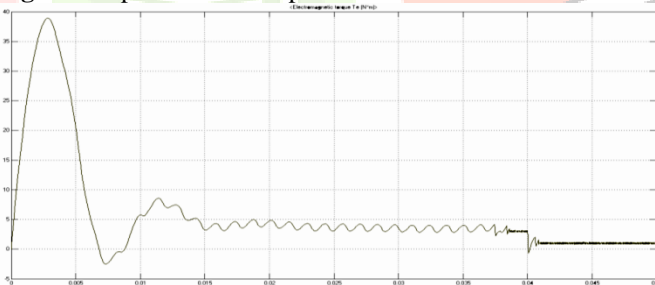


Fig. 12 Scope view of electromagnetic torque of PMSM with PID controller

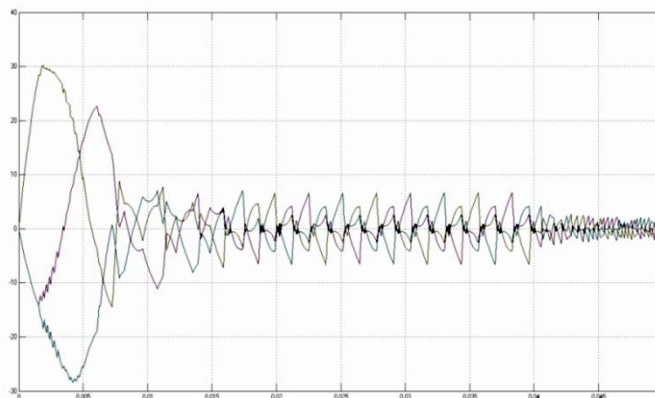


Fig. 13 Scope view of 3 phase current of PMSM with PID controller

From the above model with PID controller and its response, it is see that the raising time and settling time comes after.04 sec so here we can say that the controlling of PMSM.

7.2 Speed control model using Fuzzy Logic controller

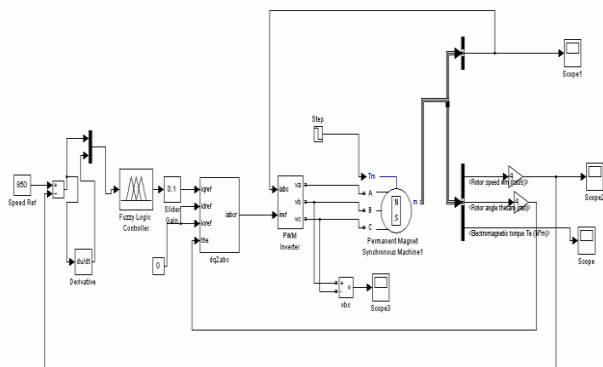


Fig. 14 MATLAB model of PMSM using Fuzzy logic controller

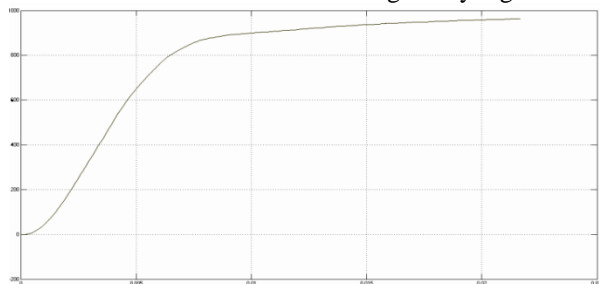


Fig. 15 Scope view of the speed of PMSM with fuzzy logic controller

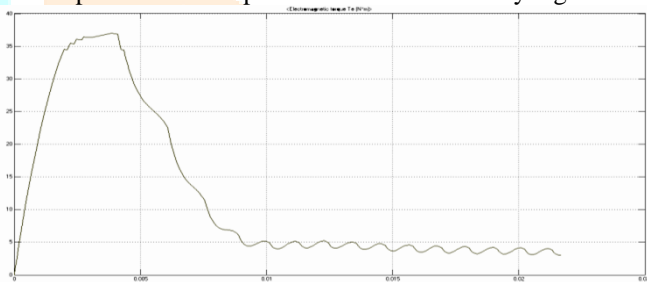


Fig. 16 Scope view of electromagnetic torque of PMSM with fuzzy logic controller



Fig. 17 Scope view of 3 phase current of PMSM with fuzzy logic controller

From the above model with fuzzy controller and its response, we can see that the rising time and settling time comes rapidly in comparison to PI and PID controllers so here we can say that the controlling speed of permanent magnet synchronous motor is much easy than the PI and PID controller.

8. CONCLUSIONS

The speed control of permanent magnet synchronous motor using fuzzy logic controller has been investigated in this work. Other conventional controllers were also used such as PI and PID. Comparison the result of speed control by all the three controllers, it is observed that fuzzy logic controller gave the best response as compared to other. The settling time of the speed was comparatively less than that of the other two controllers and it is also tabulated in 3.

Table 3: Comparison between PI and Fuzzy controller

Controller	Rise time (sec)	Settling time (sec)	Peak overshoot	Error( $e_{ss}$ ) %
PI controller	12	22	8	No Control
PID controller	0.6	4	1.32	21.27
Fuzzy controller	0.8	1.8	1.0	2.71

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